

DESIGN OF LOW POWER PULSE TRIGGERED FLIP FLOP USING CONDITIONAL PULSE-ENHANCEMENT SCHEME

L.Karthikeyan, Mr.T.Guna sekar

Abstract— Flip-flops are the major storage elements in all SOC's of digital design. They accommodate most of a power that has been applied to the chip. Flip-flop is one of a most power consumption component. It is important to reduce a power dissipation in both clock distribution networks and flip-flops. The power delay is mainly due to clock delays. The delay of a flip-flops should be minimized for efficient implementation. The concept of this project is to reduce the power consumption and to increase the speed and functionality of the chip. This project moves around in replacing conventional master-slave based flip flop to a pulse triggered flip flop which acts as a tribute alternate for low power applications. Initially in the critical path the pulse generation controls logic along with SVL function. A simple transistor SVL design is used to reduce the circuit complexity. In this scheme transistor sizes and pulse generation circuit can be further reduce for power saving. Here UMC CMOS180nm technology is use in SPICE tool to design the proposed structure. This would bring up the result in power saving approximately to 38.4%.

Index Terms— Flip-flop, low power,svl.

I. INTRODUCTION

Flip-Flops and latches are the basic elements for storing information. One latch or Flip-Flop can be store one bit information. The main difference between latches and flip-flops is that for latches, their output are constantly affected by their input as long as the enable signal asserted. In other words, when they are enabled their content change immediately when their inputs change. Flip-flops on the other hand have their content change only either at the rising or falling edge of the enable signal. This enable signal is usually the control a clock signal. After the rising or falling edge of a clock, flip-flop content remains constant even.

If the input changes. There are basically four main types of latches and flip-flops: SR,JK,Dand T.The major differences in these flip-flop [13] types are the number of inputs they have and how they

change state. For each type, there are also different variations

that enhance their operations. Figure 1(a), (b) illustrates the difference between positive edge triggered flip flop and active high latch. As it can be seen in this figure possible changes of input can be seen at the output of the latch while it is transparent.

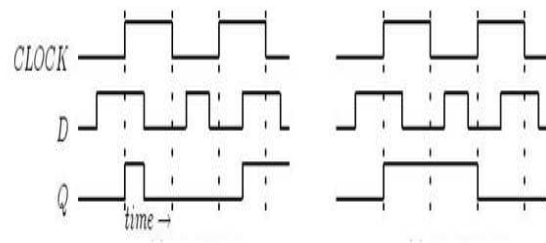
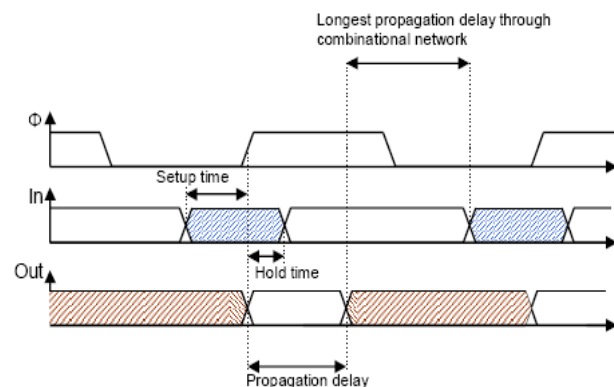


Fig 1 (a) Active High latch (b) Positive Edge Triggered Flip-Flop

The performance of a flip-flop is measured by three important timings and delays: propagation delay, setup time and hold time. They reflect in the system level performance of the Flip-Flops [5]. Setup time and hold time define the relationship between the clock and input data as shown in the Figure 1(c)

Setup time and hold time describe the timing requirements on the D input of a Flip-Flop with respect to the Clk input. Setup and hold time define window of time which the D input must be valid and stable in order to assure valid data on the Q output. Setup Time (T_{su}) – Setup time is the time that the D input must be valid before the Flip-Flop samples. Hold Time (T_{h}) – Hold time is the time that D input must be maintained valid after the Flip-Flop samples. Propagation Delay (T_{pd}) – Propagation delay is the time that takes to the sampled D input to propagate to the Q output.



Manuscript received January 03, 2014.

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Fig 1(c) Timing Diagram.

II. TECHNIQUES FOR IMPLEMENTING IMPLICIT PULSE-DATA CLOSED TO OUTPUT TRIGGERED FLIP-FLOPS

Implicit pulse generation is often considered to be more power efficient than explicit pulse generation. This is because of the former merely controls of the discharging path while the latter needs to physically generate a pulse train. Implicit-type design however, face a lengthened discharging path in latch design, which lead a inferior timing characteristics [1].

In digital circuit design, large proportion contributes to synchronous design and they are operated based on the clock signal to reduce the complexity of the circuit design. In the design of sequential circuits, major challenge is the design of an efficient D flip-flop (DFF). Several static/dynamic DFF architectures have been proposed in [1]-[10]. The topology comparison commences with the conventional single edge triggered flip-flop SET [1] typically latch data either at the positive or negative edge of the clock. A SET FF can be configured to operate as master slave latch by cascading the sequential structure but it is incompetent as half of the clock edges are wasted, while the full implementation cost the complete clock is endured. Next topology is Double Edge Triggered flip-flop DET, which can be triggered at the positive as well as the negative edges.

The implicit type flip-flop generates the pulse inside the flip-flop. The circuit diagram of ip-DCO is shown in the Figure 3.1. In ip-DCO the clock signal and complement of the clock signal generates a narrow pulse of short pulse width [6]. During this pulse the output follows the input. First, during the rising edge, of the nMOS transistors N2 and N3 are turned on. If data remain high, node x will be discharged on every rising edge of the clock. This leads to large switching power. The other problem of that node x controls two larger MOS transistors (P2 and N5). The large capacitive load to node x cause speed and power performance degradation. When the x as denoted floating node,

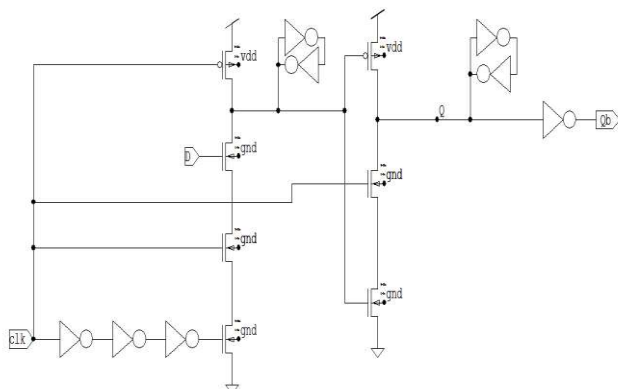


Fig. 2. ip-DCO

The node x controls two larger transistors P2 and N5, this leads to large capacitive load to node x causes power performance degradation. Fig. 3 shows the simulation

waveforms of the proposed P-FF design against the MHLLF design.

A. SCCER

In this design shown in fig 3, a weak pull up transistor P1 is employed in conjunction with an inverter I2 to reduce the load capacitance of node. The discharge path contains NMOS transistors N2 and N1 connected in series. In order to eliminate to superfluous switching node X, an extra NMOS transistor N3 is employed. Since N3 is controlled by Q_fdbk no discharge occurs if input data remains high.

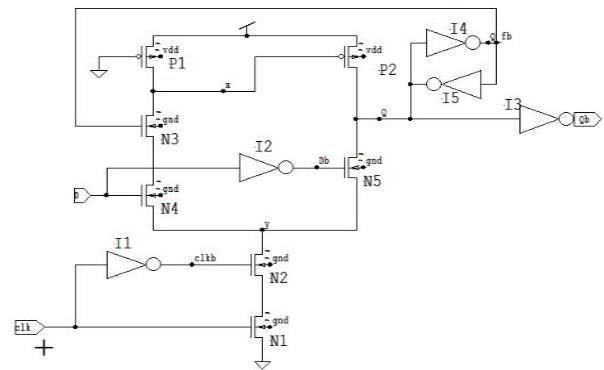


Fig 3. SCCER

The worst case timing of this design occurs when input data is "1" and node X is discharged through four transistors in series, i.e., N1 through N4, while combating with the pull up transistor P1.

B. ATSVL with SCCER

In order to avoid that we can incorporate the leakage reduction circuit called "Self-Controllable Voltage Level Circuit" to this flip flop to reduce the power consumption. This SVL [11] method is implemented in memory circuits in prior papers to reduce the power consumption. In this paper the same SVL approach is applied to the new flip flop to reduce the

leakage current and power which leads to total power reduction. Two blocks, Upper SVL and Lower SVL circuit will give the maximum V_{dd} and minimum $ground$ level to the flip flop(load) when *active* mode. In other hand it will give lower V_{dd} and higher $ground$ level to the load in *standby* mode. The block diagram of SVL-flip flop is shown in figure 4. The two circuits called upper SVL (U-SVL) and Lower SVL (L-SVL) is used to construct the above fourth proposed flip flop. Upper SVL consists of one PMOS (pSW) act as a switch and multiple NMOS (nRSm) act as resistors connected in series. Similarly, Lower SVL constructed by one NMOS (nSW) and multiple PMOS (pRSm) in series.

$clk=1$ and $clkb=0$, P1 and N4 is ON but N1 and P4 is OFF. Therefore the Upper and Lower SVL blocks can supply a maximum supply voltage V_{dd} and a minimum $ground$ level V_{ss} respectively. While the stand-by mode i.e $clk=0$ and $clkb=1$, P1 and N4 is OFF but N1 and P4 is ON. Upper SVL circuit generate lower supply $V_{dd} (=V_{dd}-V_n < V_{dd})$ to a flip flop and Lower SVL circuit gives higher ground level voltage $V_{ss} (=V_p > 0)$. Where V_n and V_p is the total voltage drop of N2, N3 and P2, P3 respectively. In this mode the back-gate bias (V_{BGS}) of the P3 and N6 are increased. Then V_{ts} of P1 and N4 also increases. Thus, the leakage current and power is

decreases. Finally the total power consumption of flip flop is reduced.

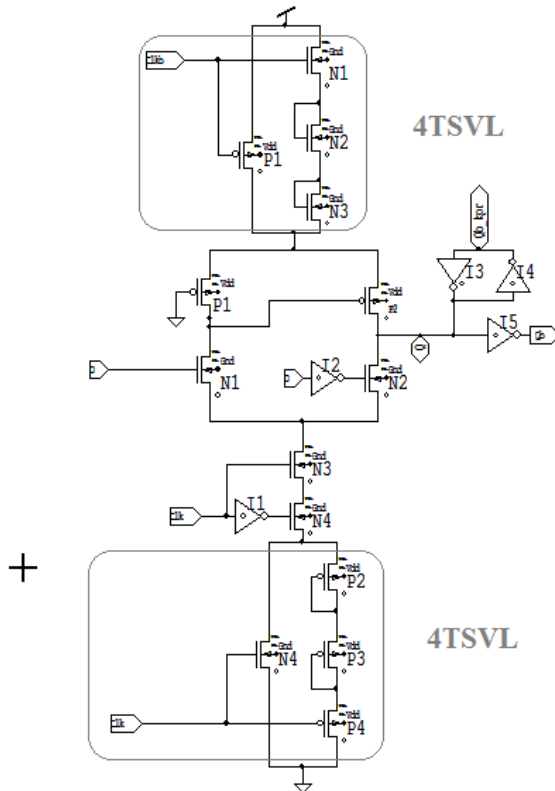


Fig 4. 4TSVL logic with SCCER

III. SIMULATION RESULTS

The simulation results were obtained from TSPICE in 180nm CMOS process at room temperature VDD is 1V. All flip flops were simulated with output load capacitance *Cload* and layout level. Following flip flop metrics are carried out to compare the performance flip flops.

Total no of Transistors: The total number of transistors is measured which contribute more area and power consumption in the integrated circuit design.

Power: It is the total power consumption of flip flop in terms of μw (micro watt).the maximum power denotes the maximum power consumption of flip flop. The minimum power describe the minimum power requirement to trigger the flip flop

To illustrate the merits of the presented work, Fig. 5 shows the simulation waveforms by using T-Spice tool. Fig 5(a) describe the output waveform obtain for IPDCO .Fig 5(b) describes the output curve of SCCER flip flop.Fig5(c) shows the performance curve of proposed flip flop with SVL logic.

Table I summarizes some important performance indexes of these P-FF design. These include transistor count, average power, maximum power, minimum power and the simulation time.

Fig 6 shows the comparison chart of three flip flops which denotes the average power, maximum power and the minimum power of ipdco, sccer and svl logic.

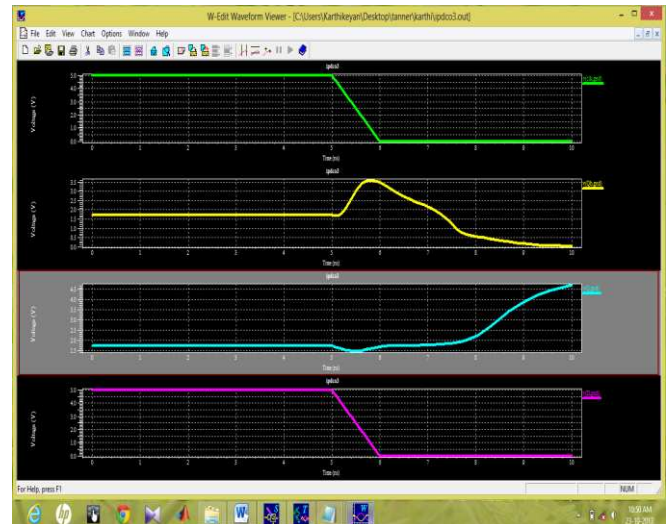


Fig.5(a).Output waveform of IPDCO

When the data is apply '1' at the time of clock as rising edge, then the output of Q is '1' because it act as the D flip flop operations.

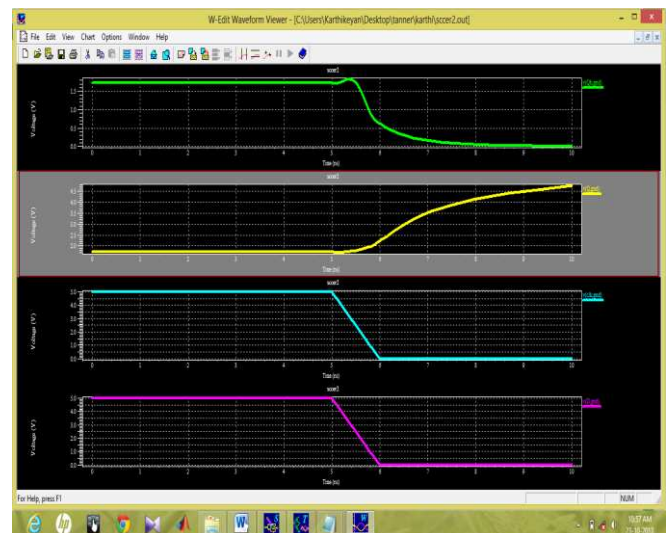


Fig.5 (b).Output waveform of SCCER

The waveform of SCCER is shown in the Figure 5 (b).In this circuit consist of input signal for clock and Data, output for Q and Qb.

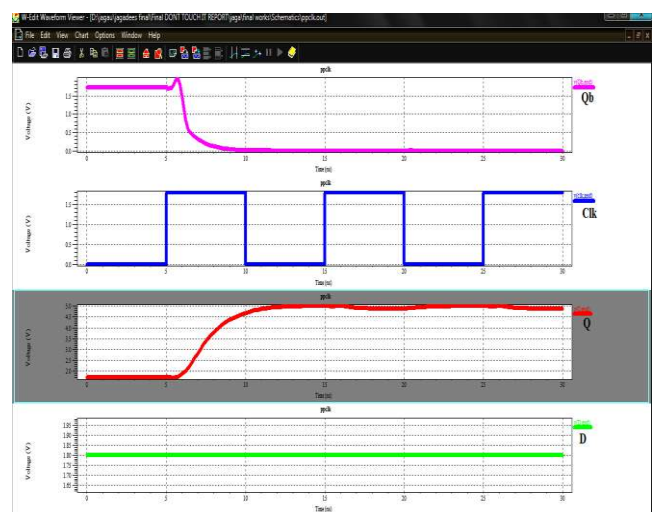


Fig.5(c).Output waveform of SVL logic

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When the fig 5(c) shows that operation of the SVL logic output and performed D flip flop, initially $clk=0$, whatever input does not consider the output '0'

Table I comparison table

P-FF	Ip-DCO	MHLFF	SCCER	4TSVL LOGIC
Number of transistors	23	19	17	24
Average Power(w)	1.141	1.029	0.691	0.501
Max.Power (μW)	1.218	0.923	0.423	0.275
Min.Power (μW)	0.2903	0.2383	0.2067	0.122

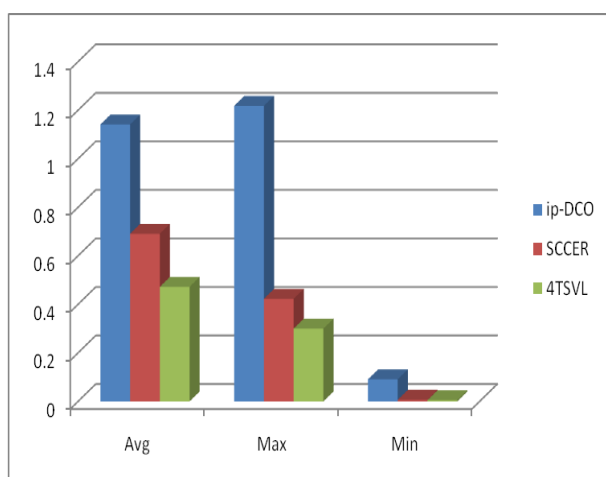


Fig.6.comparison chart

IV. CONCLUSION

This project, the various Flip flop design like, ip-DCO, MHLFF, SCCER and Proposed flip-flop design with pulse control scheme are discussed. These flip-flop were been designed in Tanner tool and the results and waveforms are also obtained. The comparison table also added to verify the designed methods. Even though, when the circuit is in idle condition the circuit consumer some power which is said to be leakage power. A master slave flip flop is initially replaced with pulse triggered flip flop and a simple SVL logic is used to reduce the power consumption and increase the device functionality. The result and the comparison chart describe that the SVL logic increased the performance of flip flop.

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